



Millimeter-Wave MIMO Architectures for 5G Gigabit Wireless

GLOBECOM Workshop on
Emerging Technologies for 5G Wireless Cellular Networks
December 8, 2014

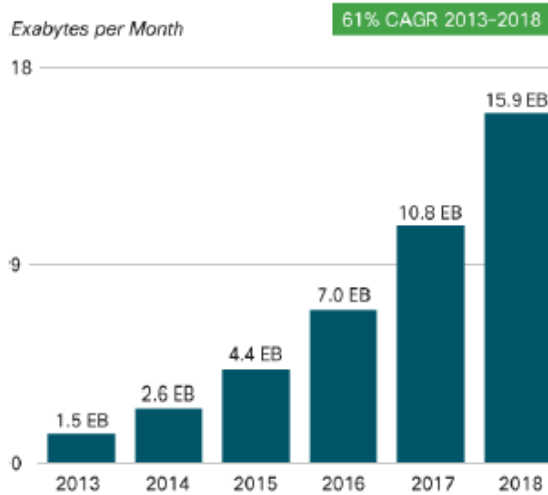
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<http://dune.ece.wisc.edu>

Supported by the NSF and the Wisconsin Alumni Research Foundation

Explosive Growth in Wireless Traffic

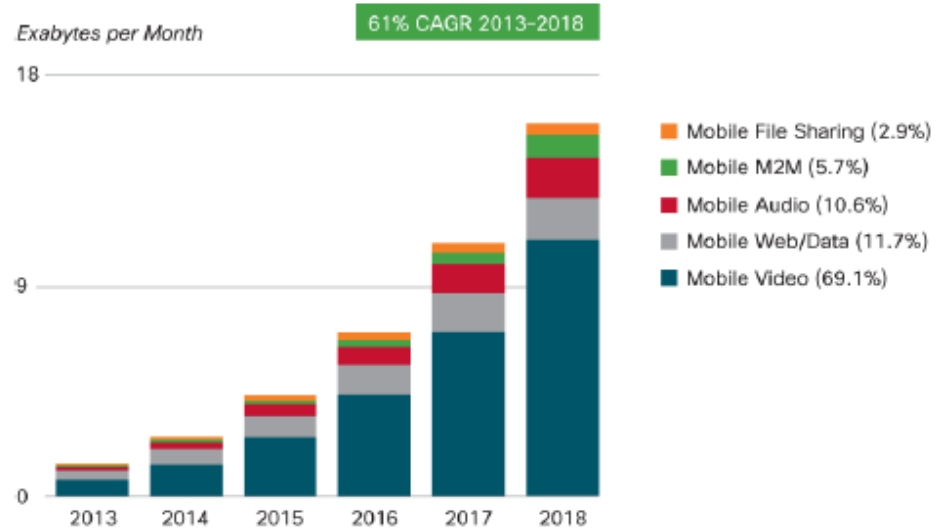
Figure 1. Cisco Forecasts 15.9 Exabytes per Month of Mobile Data Traffic by 2018



Source: Cisco VNI Mobile, 2014

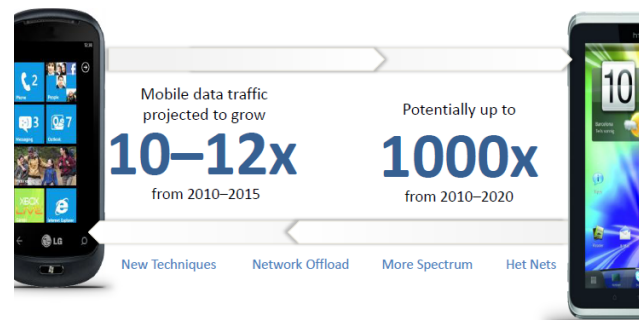
(2014 Cisco visual networking index)

Figure 10. Mobile Video Will Generate Over 69 Percent of Mobile Data Traffic by 2018



Figures in parentheses refer to traffic share in 2018.

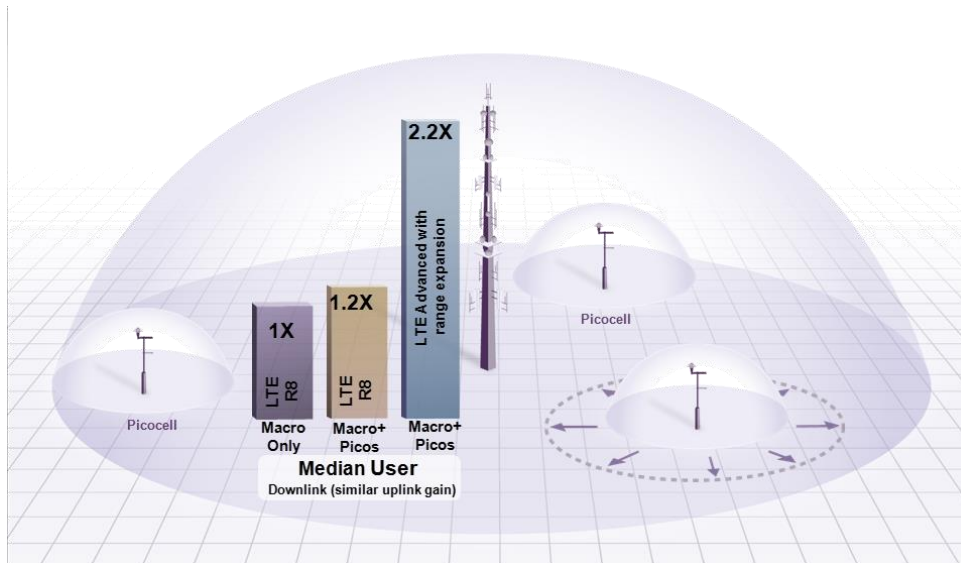
Source: Cisco VNI Mobile, 2014



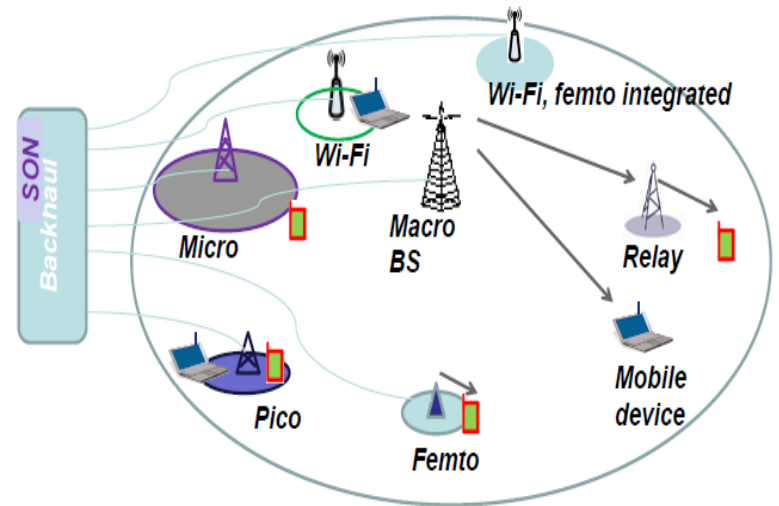
(Qualcomm)

Current Industry Approach: Small Cells & Heterogeneous Networks

Key Idea:
Denser spatial reuse
of limited spectrum



Courtesy: Dr. T. Kadous (Qualcomm)



Courtesy: Dr. J. Zhang (Samsung)

Some challenges: **interference, backhaul**

New Opportunity: mm-wave Band



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UNITEI 3kHz

STAT 300kHz
FREQUENCY

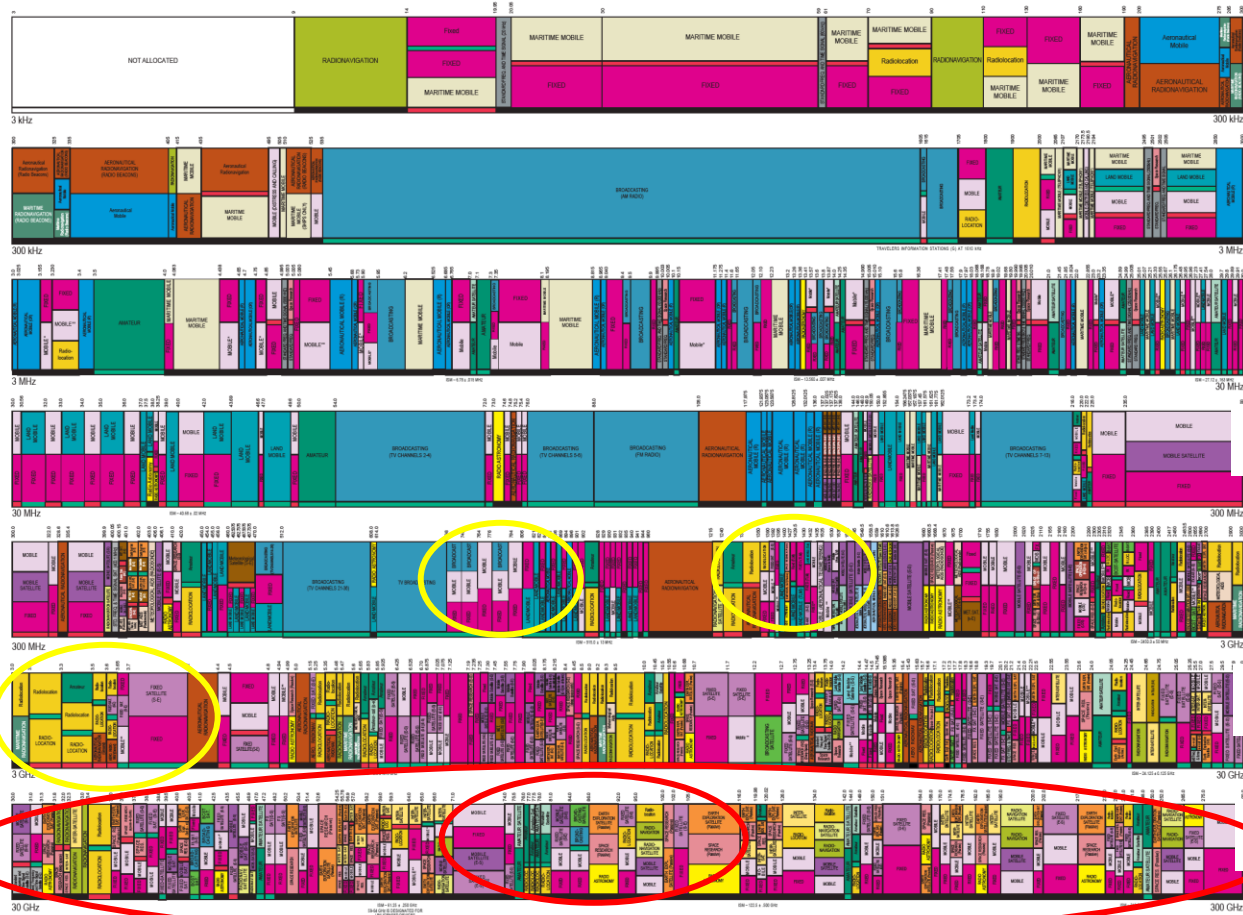
ALLOCA 3MHz
THE RADIO SPECTRUM

RADIO SERVICES COLOR LEGE 30MHz

300MHz

3GHz

30GHz



300kHz

3MHz

30MHz

300MHz

3GHz

30GHz

300GHz

ACTIVITY CODE
 GOVERNMENT EXCLUSIVE
 NON-GOVERNMENT EXCLUSIVE
 ALLOCATION USAGE DESIGNATI
 SOURCE: EXAMPLE DESCRIPTION
 Primary: FIXED Capital Letters
 Secondary: Mobile 1st Capital with lower case letters

U.S. DEPARTMENT OF COMMERCE
 National Telecommunications and Information Administration
 Office of Spectrum Management
 October 2003

Current cellular wireless: 300MHz - 5GHz

Mm-wave - Short range: 60GHz

Long range: 30-40GHz, 70/80/90GHz

Mm-wave Wireless: 30-300 GHz

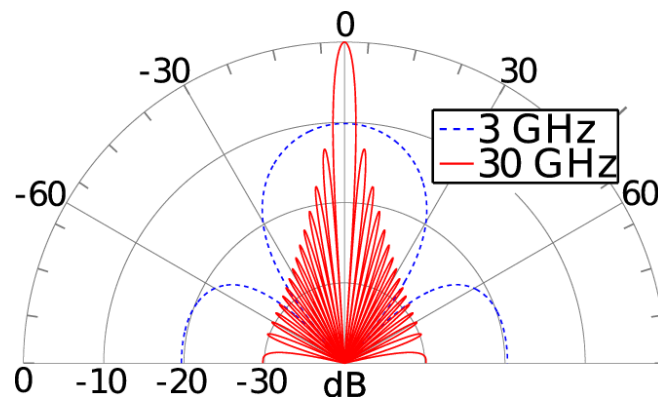
A unique opportunity for addressing the wireless data challenge

- Large bandwidths (GHz)
- High spatial dimension: short wavelength (1-100mm)

Compact high-dimensional multi-antenna arrays

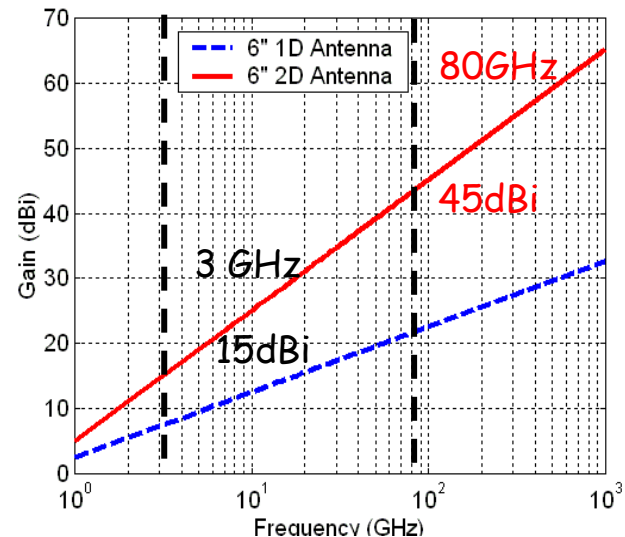
6" antenna: 6400-element antenna array (80GHz)

Highly directive narrow beams
(low interference/higher security)



Beamwidth: 35 deg @ 3GHz 2 deg @ 80GHz

Large antenna gain



Current & Emerging Applications

- Wireless backhaul; alternative to fiber
- Indoor wireless links (e.g., HDTV) IEEE 802.11ad, WiGig
- Smart base-stations for 5G mobile wireless (small cells)
- New cellular/mesh/heterogeneous network architectures
- Space-ground or aircraft-satellite links

Multi-Gigabits/s speeds
Multiple Beams

Key Opportunities and Challenges

Key Operational Functionality:

Electronic multi-beam steering & MIMO data multiplexing

Key Challenges:

- Hardware complexity: spatial analog-digital interface
- Computational complexity: high-dimensional DSP

Our Approach: **Beamspace MIMO**



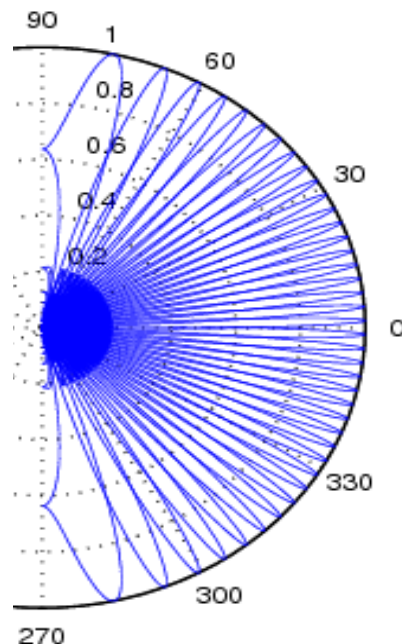
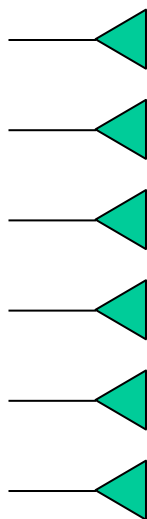
Beamspace MIMO

Multiplexing data into
multiple highly-directional and high-gain beams

Discrete Fourier Transform
(DFT)

Antenna space
multiplexing

n-element array
($\frac{\lambda}{2}$ spacing)



Beamspace
multiplexing

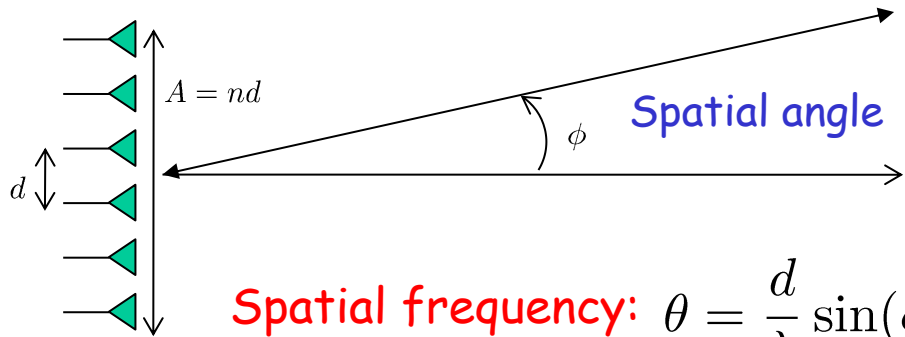
n orthogonal beams

spatial channels

n dimensional
signal space

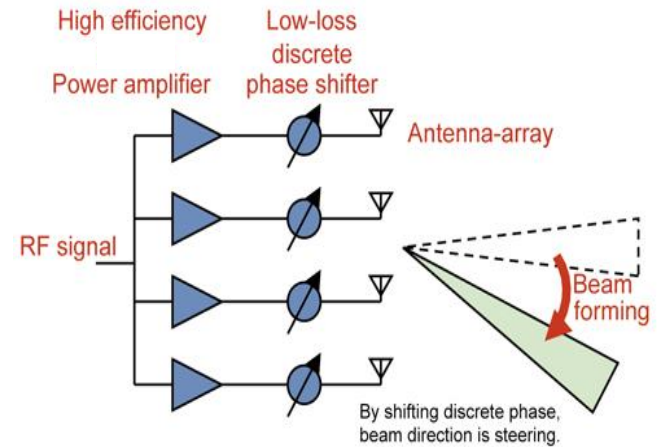
(AS'02, AS&NB '10, JB, AS, & NB '14)

n-element Antenna (Phased) Array



Spatial frequency: $\theta = \frac{d}{\lambda} \sin(\phi)$

$$-\frac{\pi}{2} \leq \phi \leq \frac{\pi}{2} \iff d = \frac{\lambda}{2} \iff -\frac{1}{2} \leq \theta \leq \frac{1}{2}$$



TX: steering vector
or
RX: response vector

$$\mathbf{a}_n(\theta) = \begin{bmatrix} 1 \\ e^{-j2\pi\theta} \\ \vdots \\ e^{-j2\pi\theta(n-1)} \end{bmatrix}$$

n-dimensional
spatial sinusoid

Orthogonal Spatial Beams

Spatial resolution/beamwidth:

$$\Delta\theta_o = \frac{1}{n} \longleftrightarrow \Delta\phi_o = \frac{\lambda_c}{A}$$

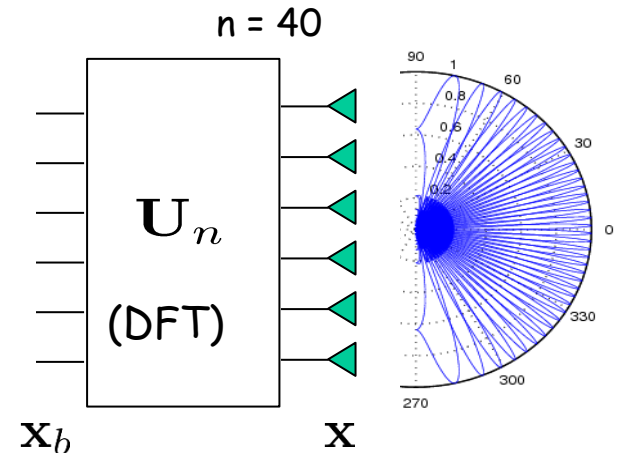
n orthogonal spatial beams

$$\theta_i = i\Delta\theta_o = \frac{i}{n} \quad i = 0, \dots, n-1$$

DFT spatial modulation matrix:

$$\mathbf{U}_n = \frac{1}{\sqrt{n}} [\mathbf{a}_n(\theta_0), \mathbf{a}_n(\theta_1), \dots, \mathbf{a}_n(\theta_{n-1})]$$

(n -dimensional orthogonal basis)



$$\mathbf{x}_b = \mathbf{U}_n^H \mathbf{x}$$

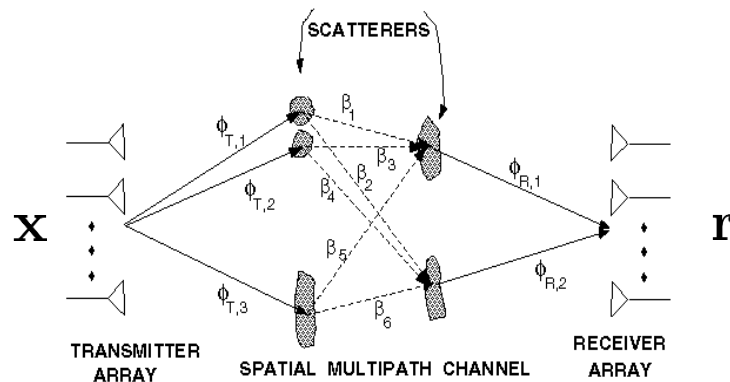
$$\mathbf{x} = \mathbf{U}_n \mathbf{x}_b$$

$$\mathbf{U}_n^H \mathbf{U}_n = \mathbf{U}_n \mathbf{U}_n^H = \mathbf{I}_n$$

Unitary

Antenna vs Beam-space Representation

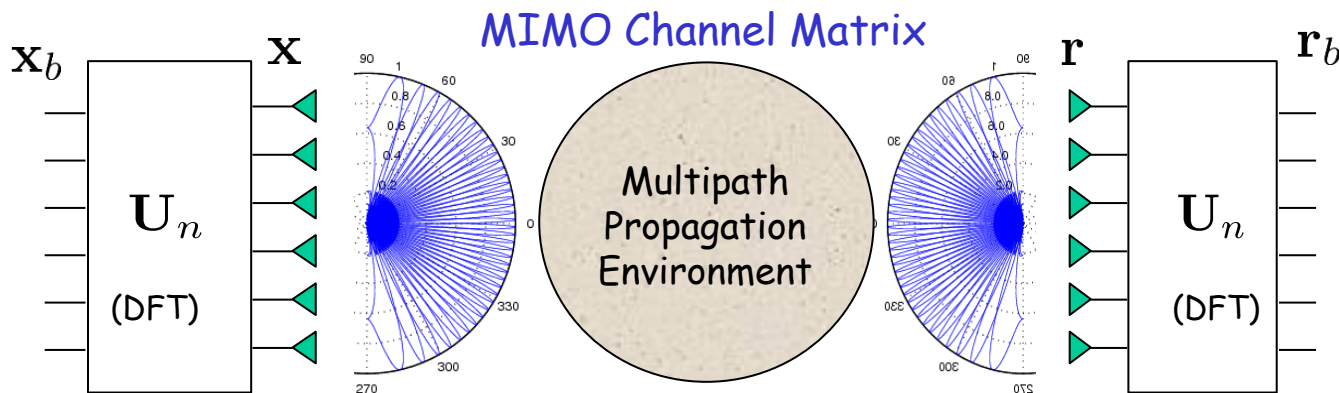
$n \times n$
MIMO system



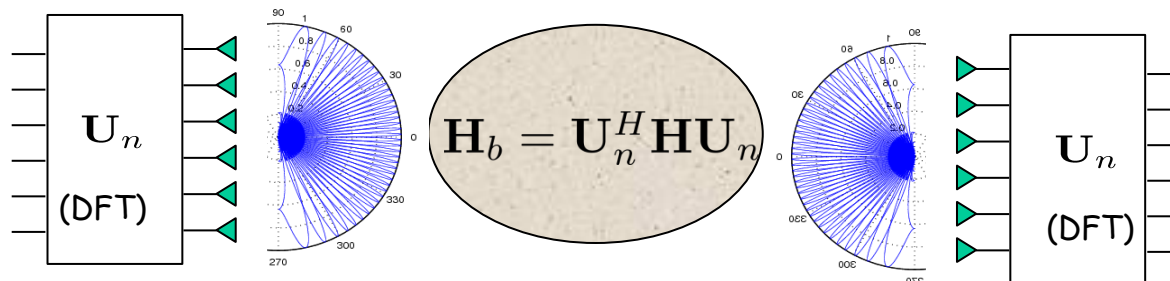
Antenna domain: $\mathbf{r} = \mathbf{H}\mathbf{x} + \mathbf{w}$

Beam domain: $\mathbf{r}_b = \mathbf{H}_b\mathbf{x}_b + \mathbf{w}_b$

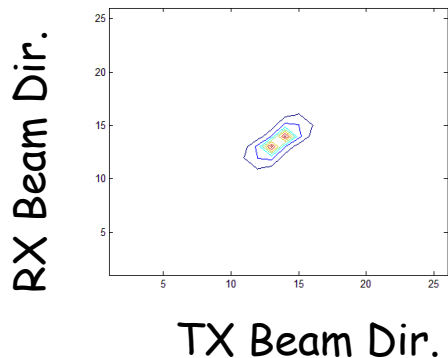
TX: $\mathbf{x} = \mathbf{U}_n\mathbf{x}_b$ $\mathbf{H}_b = \mathbf{U}_n^H \mathbf{H} \mathbf{U}_n$ RX: $\mathbf{r}_b = \mathbf{U}_n^H \mathbf{r}$



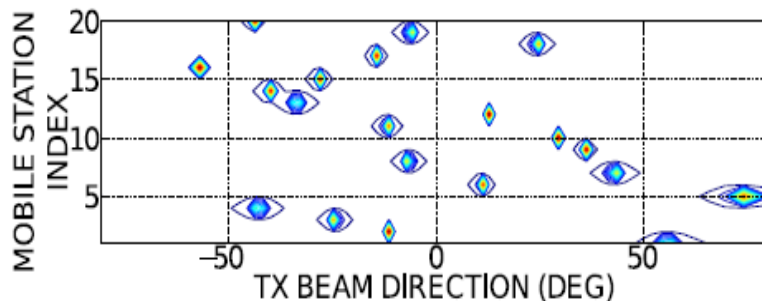
Key Characteristic at mmW: Beamspace Channel Sparsity



Point-to-point



Point-to-multipoint



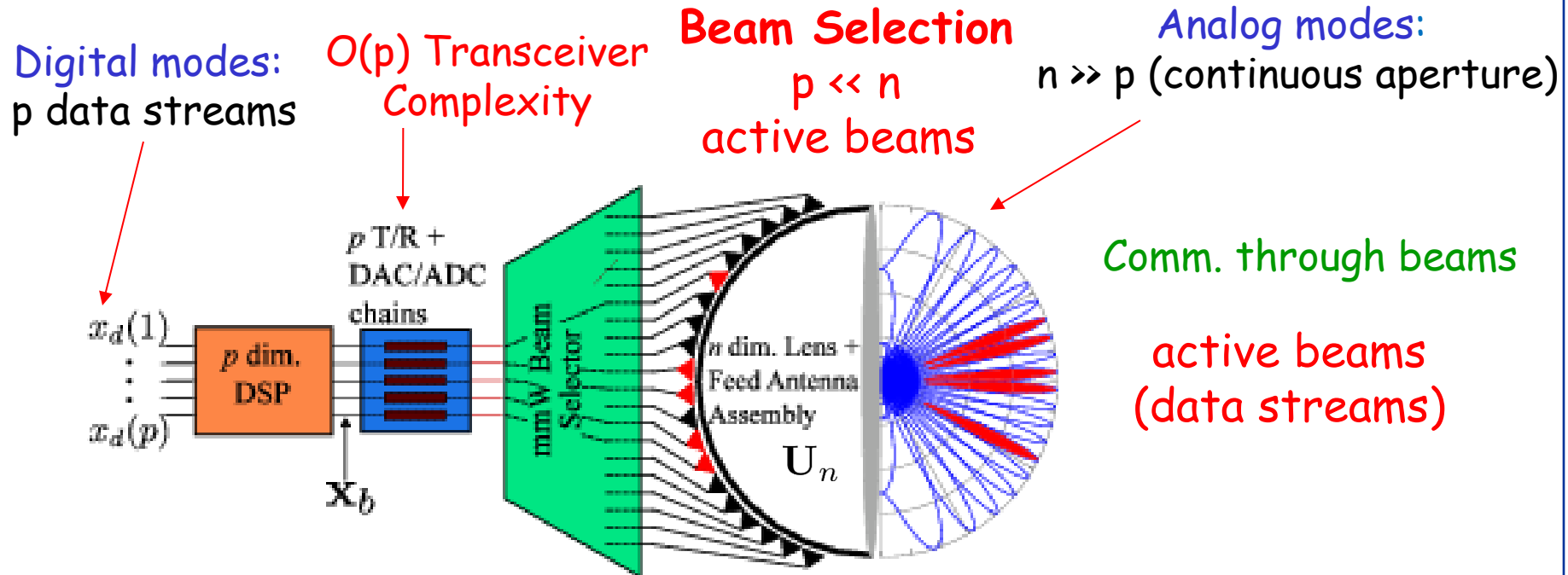
- Directional, quasi-optical
- Primarily line-of-sight
- Single-bounce multipath

Communication occurs in a low-dimensional (p) subspace
of the high-dimensional (n) spatial signal space

How to optimally access the communication subspace
with the lowest - $O(p)$ - transceiver complexity?

Continuous Aperture Phased (CAP) MIMO

Practical Hybrid Analog-Digital Transceiver Architecture
for BeamSpace MIMO (patented)

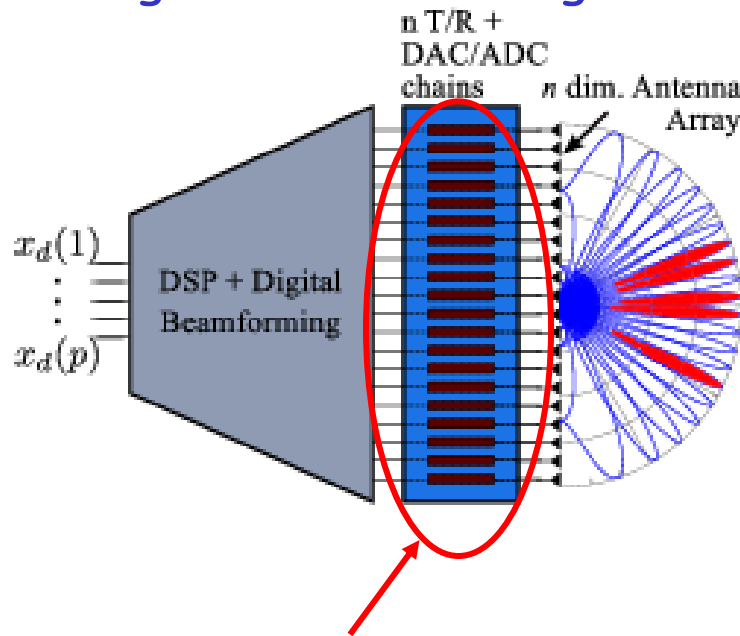


Lens computes analog spatial DFT:
direct access to beamspace

Performance-Complexity Optimization:
Optimal Performance with Lowest Hardware Complexity

Spatial Analog-Digital Interface: Digital vs Analog Beamforming

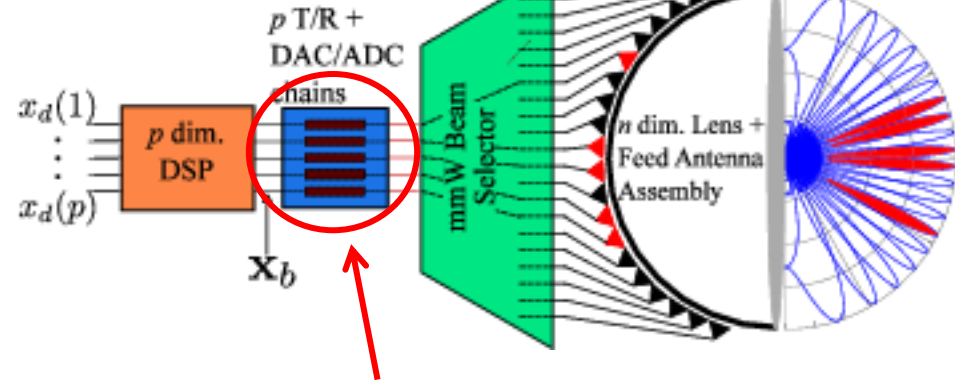
Conventional MIMO:
Digital Beamforming



$O(n)$ transceiver complexity

Beam Selection
 $p \ll n$
active beams

CAP MIMO:
Analog Beamforming



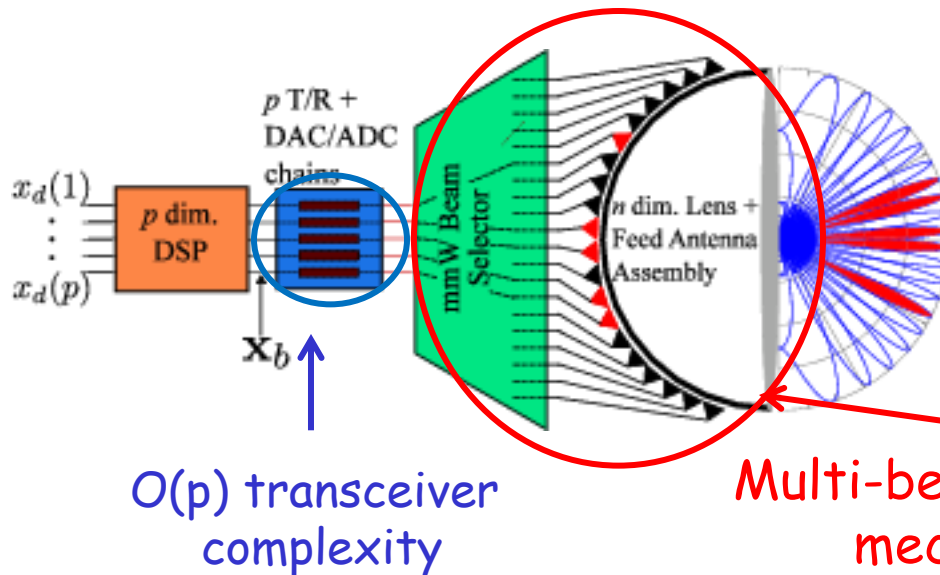
$O(p)$ transceiver complexity

n : # of conventional MIMO array elements (1000-100,000)

p : # spatial channels/data streams (10-100)

CAP-MIMO vs Phased-Array-Based Architectures for Analog Beamforming

CAP-MIMO:

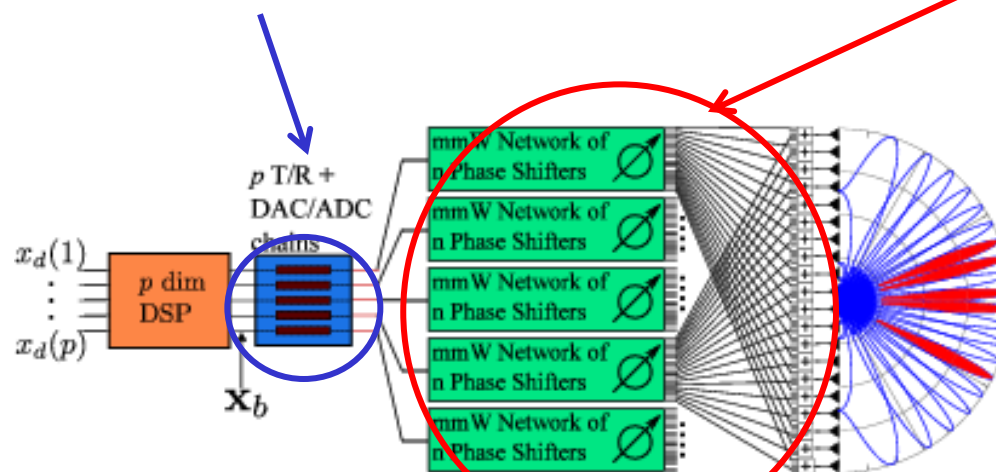
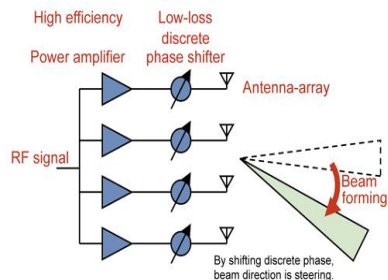


Lens
+
Beamspace Array
+
mmW Beam Selector Network

$O(p)$ transceiver complexity

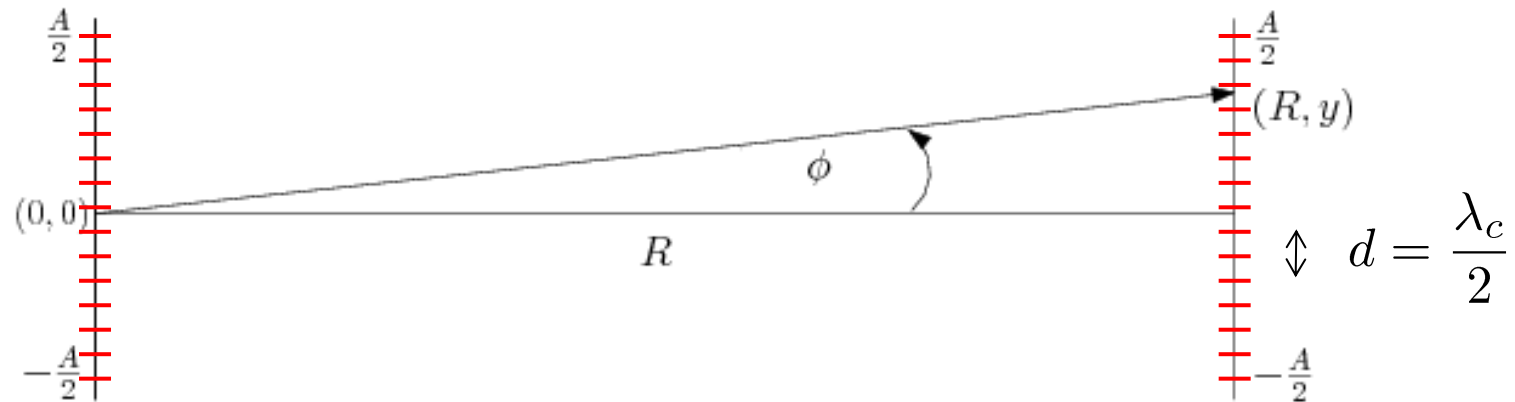
Multi-beam forming mechanism

Phased-Array Based Architecture:



Phase Shifter Network (np)
+
Combiner Network

Case Study: Line-of-Sight Link



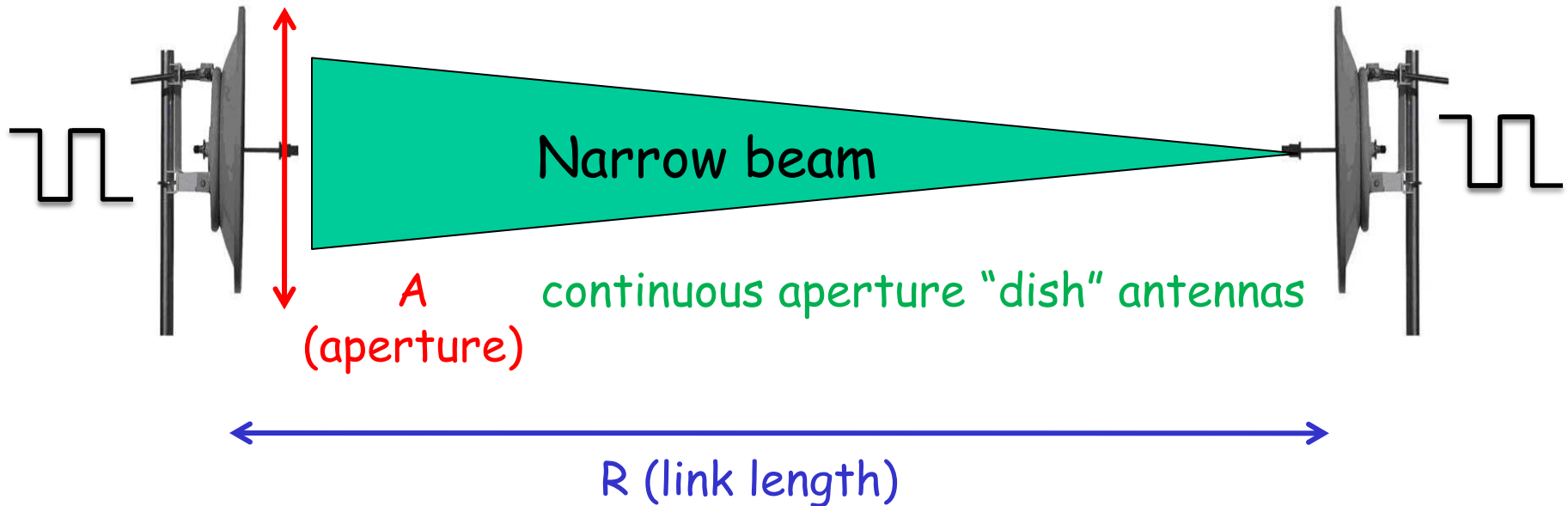
analog spatial modes: $n \approx \frac{A}{d} = \frac{2A}{\lambda_c}$ (\sim antenna/array gain)

Critical sampling of aperture $\implies n \times n$ MIMO system

Spatial domain representation: $\mathbf{r} = \mathbf{H}\mathbf{x} + \mathbf{w}$

Beamspace representation: $\mathbf{r}_b = \mathbf{H}_b\mathbf{x}_b + \mathbf{w}_b$

State-of-the-Art 1: DISH System



Pros: Large antenna gain (SNR gain) $G \propto \left(\frac{A}{\lambda}\right)^2$

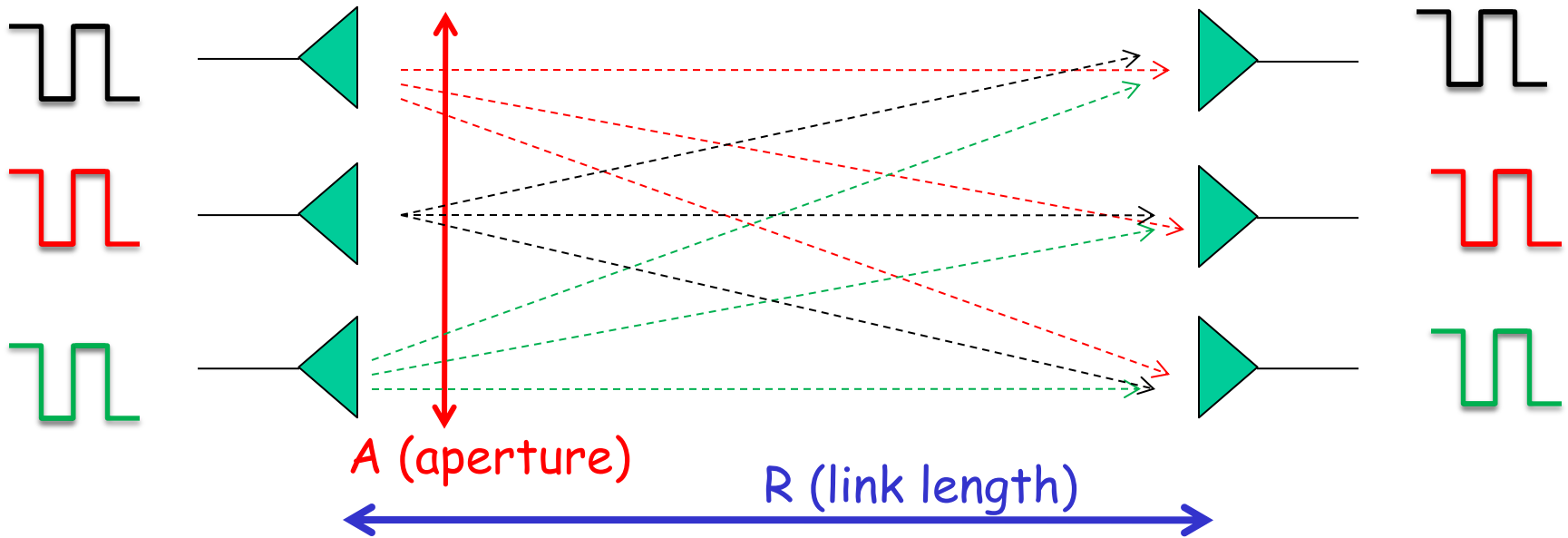
Narrow beam (continuous aperture)

Cons: Single data stream



State-of-the-Art 2: MIMO System

Discrete Antenna Arrays (wide Rayleigh spacing)



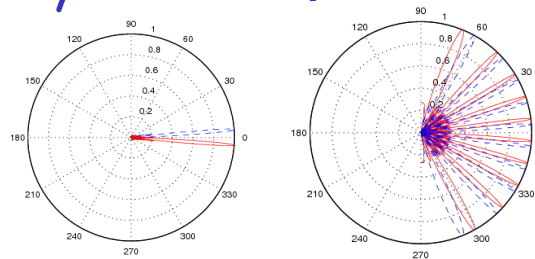
Pros: Multiplexing gain: Multiple (p) data streams
(p limited by A and R)

Madhow et. al. 06'

Bohagen et. al. 07'

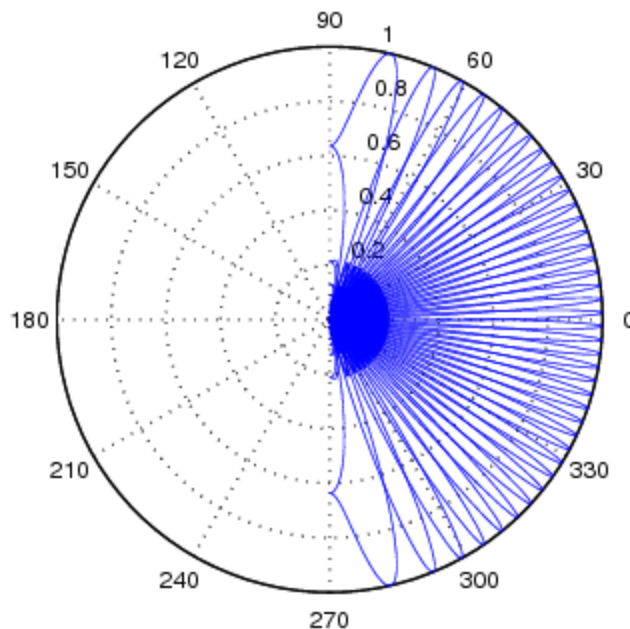
Cons: Reduced SNR gain
Grating lobes

Chalmers

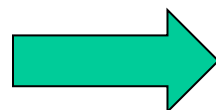
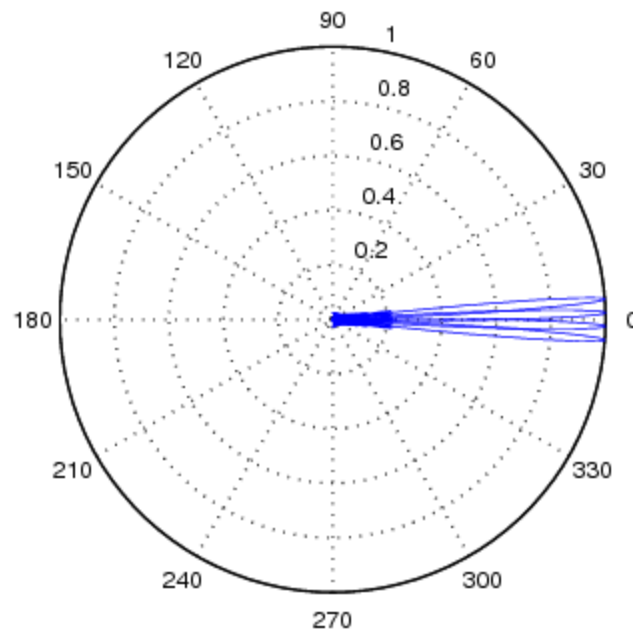


Beamspace MIMO: Coupled Orthogonal Beams

$n = 40$ orthogonal beams



$p_{max} = 4$ coupled beams



Finite
Receiver
Aperture

number of
coupled beams:
(channel rank)

$$p_{max} = \frac{2\theta_{max}}{\Delta\theta_o} = 2\theta_{max}n = \frac{A^2}{R\lambda_c}$$

(Fresnel number)

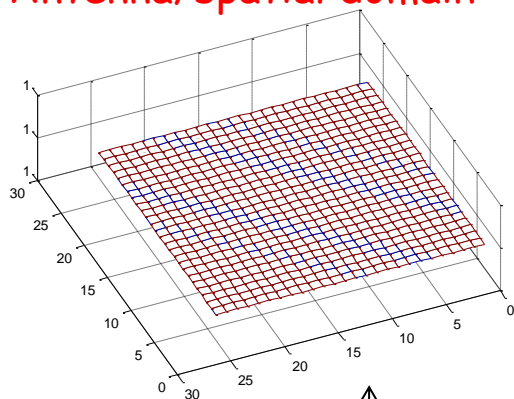
Spatial BW: $2\theta_{max} = \sin(\phi_{max}) \approx \frac{A}{2R}$ Spatial Res.: $\Delta\theta_o = \frac{1}{n} = \frac{\lambda_c}{2A}$

Near-Optimality of Beam-space MIMO

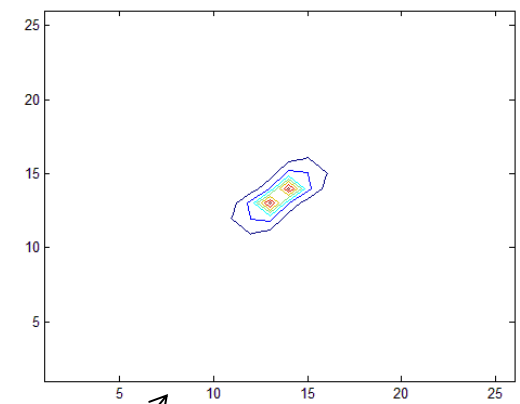
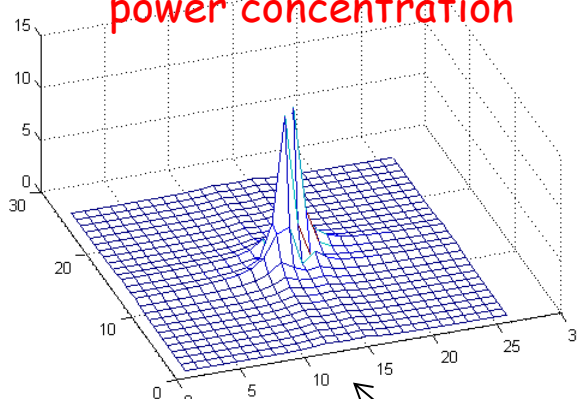
Coupled beams \sim channel eigenvectors

$$n = 26; p_{max} = 2$$

Antenna/spatial domain



Beamspace:
power concentration



$$\mathbf{H} = \mathbf{U}_n \mathbf{H}_b \mathbf{U}_n^H \iff \mathbf{H}_b = \mathbf{U}_n^H \mathbf{H} \mathbf{U}_n$$

$$\mathbf{H} = \mathbf{U}_n \mathbf{H}_b \mathbf{U}_n^H \approx \tilde{\mathbf{U}}_n (n \times p_{max}) \tilde{\mathbf{H}}_b (p_{max} \times p_{max}) \tilde{\mathbf{U}}_n^H (p_{max} \times n)$$

\uparrow approx. eigenvectors (RX subspace)
 \uparrow approx. eigenvalues (diagonal)
 \uparrow approx. eigenvectors (TX subspace)

Subspace determination through beamspace thresholding!

Capacity Comparison with State of the Art

Spectral Efficiency (bits/s/Hz)

CAP-MIMO:

Max. multiplexing gain & Max. SNR gain

$$C_{cap-mimo}(\rho) \approx p_{max} \log \left(1 + \rho \frac{n^2}{p_{max}^2} \right)$$

**MUX gain
over DISH:**

p_{max}

**SNR gain over
widely spaced MIMO:**

$$G = \frac{n^2}{p_{max}^2}$$

$$C_{dish}(\rho) \approx \log \left(1 + \rho \frac{n^2}{p_{max}^2} \right)$$

DISH

no multiplexing gain
Max. SNR gain

$$C_{mimo}(\rho) = p_{max} \log (1 + \rho)$$

Conv. (widely spaced) MIMO

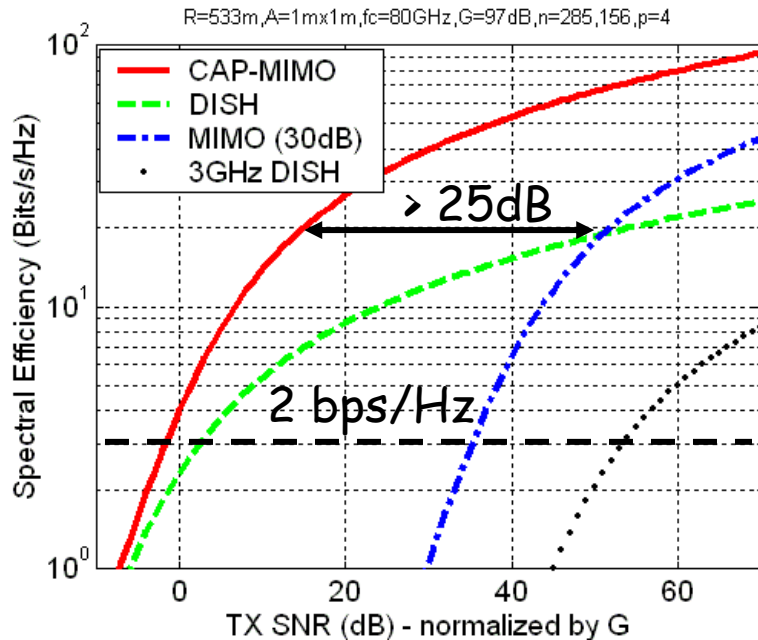
Max. multiplexing gain
small SNR gain

Potential Gains: Backhaul and Indoor Links

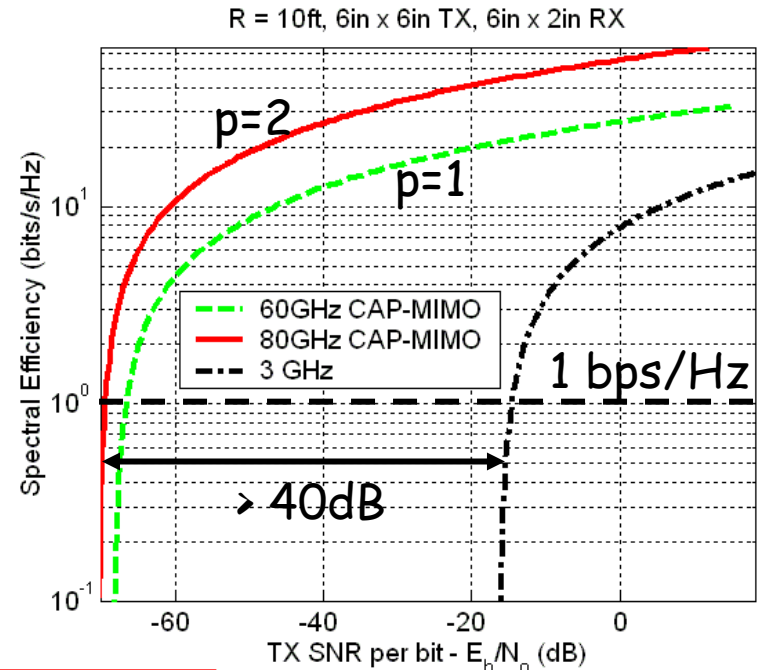


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Longer (backhaul) link:
R=533m, A=1m x 1m, 80GHz



Shorter (indoor) link:
R=10ft, TX: 6in x 6in, RX: 6in x 2in



Spectral efficiency (bits/s/Hz):

$n \sim 300,000$
 $G \sim 100\text{dB}$
 $p = 4$

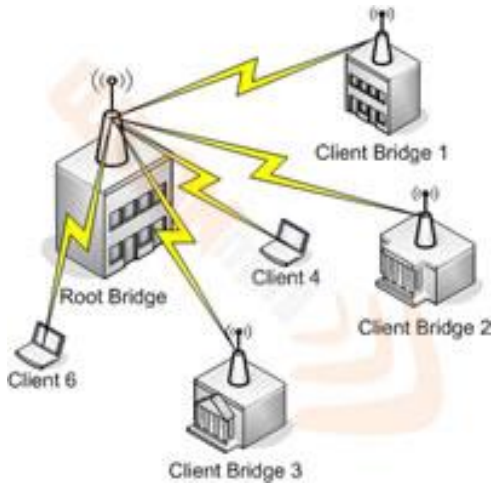
$$C(\rho) \approx p \log \left(1 + \rho \frac{n^2}{p^2} \right)$$

ρ : SNR n : spatial dimension
 p : # data streams

3GHz
 $n \sim 9, G \sim 15\text{dB}$
60GHz
 $n \sim 3000, G \sim 66\text{dB}$
80GHz
 $n \sim 6000, G \sim 66\text{dB}_{21}$

Point-to-Multipoint Links

Fixed (backhaul) and dynamic (access) links



Downlink precoding: $\mathbf{r} = \mathbf{H}^H \mathbf{x} + \mathbf{w} = \mathbf{H}^H \mathbf{G} \mathbf{s} + \mathbf{w}$

Beamspace precoding: $\mathbf{r} = \mathbf{H}_b^H \mathbf{G}_b \mathbf{s}_b + \mathbf{w}$

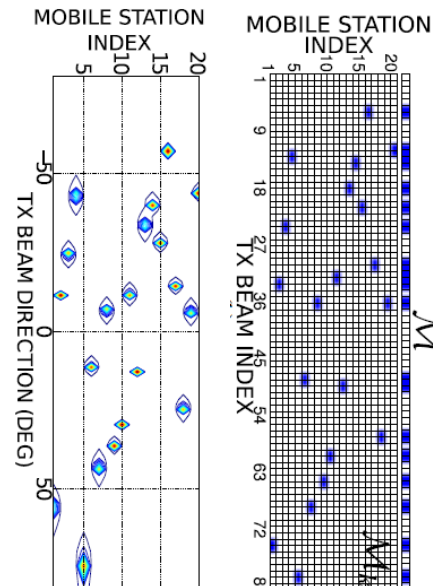
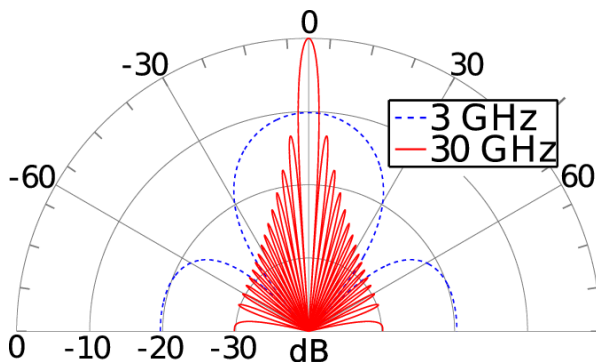
$$\mathbf{G}_b = \mathbf{U}_n^H \mathbf{G} = [\mathbf{g}_{b,1}, \mathbf{g}_{b,2}, \dots, \mathbf{g}_{b,K}]$$

Multuser channel: $\mathbf{H} = [\mathbf{h}_1, \dots, \mathbf{h}_K]$, $\mathbf{h}_k = \beta_k \mathbf{a}_n(\theta_k)$

Beamspace channel: $\mathbf{H}_b = \mathbf{U}_n^H \mathbf{H} = [\mathbf{h}_{b,1}, \dots, \mathbf{h}_{b,K}]$

Sparse Beamspace Channel

(Driverlayer.com)



Lower-dimensional system

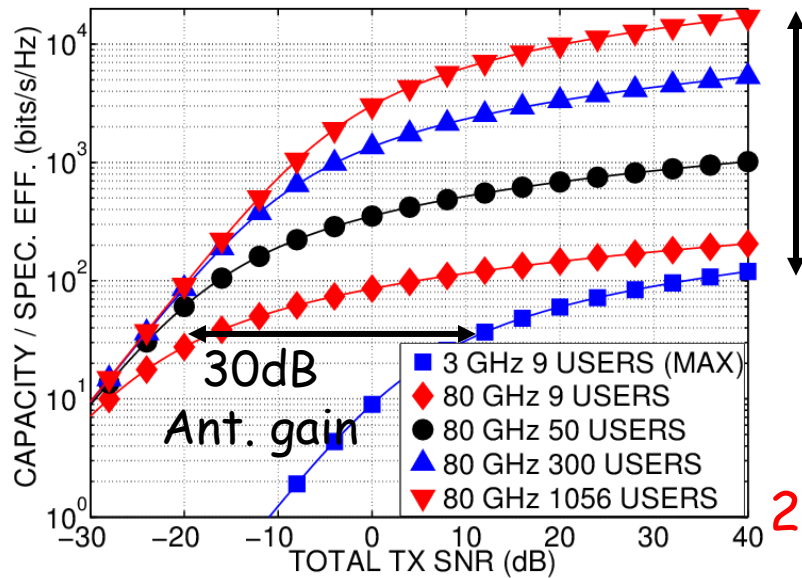
$$\mathbf{r} = \tilde{\mathbf{H}}_b^H \tilde{\mathbf{G}}_b \mathbf{s}_b + \mathbf{w}$$

$$\tilde{\mathbf{H}}_b = [\mathbf{H}_b(l, :)]_{l \in \mathcal{M}}$$

Dense BeamSpace Multiplexing

Key application: **small-cell access points**

Idealized upper bound (non-interfering users) $C_{ub}(\rho, K, n) = K \log_2 \left(1 + \rho \frac{n}{K} \right)$

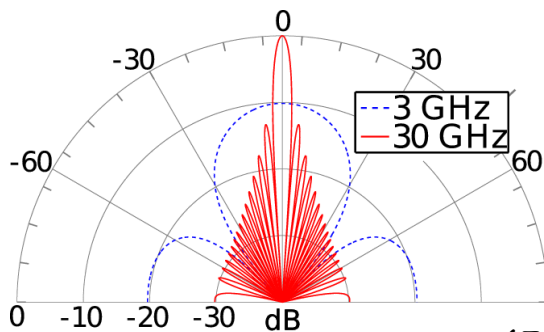


x 2-200 increase in capacity
due to **beamSpace multiplexing**

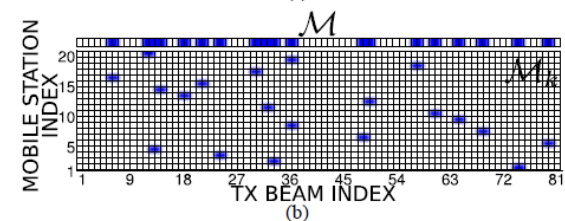
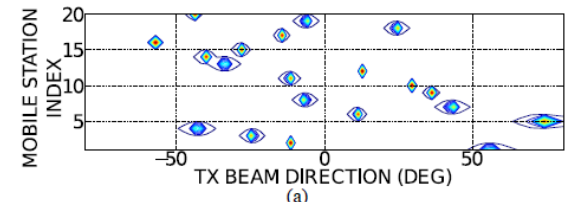
x 10-100 increase in capacity
due to **extra bandwidth**
(~1-10GHz vs 100MHz)

200Gbps-200Tbps (per cell throughput)
(20-200Gbps/user)

6" x 6" antenna

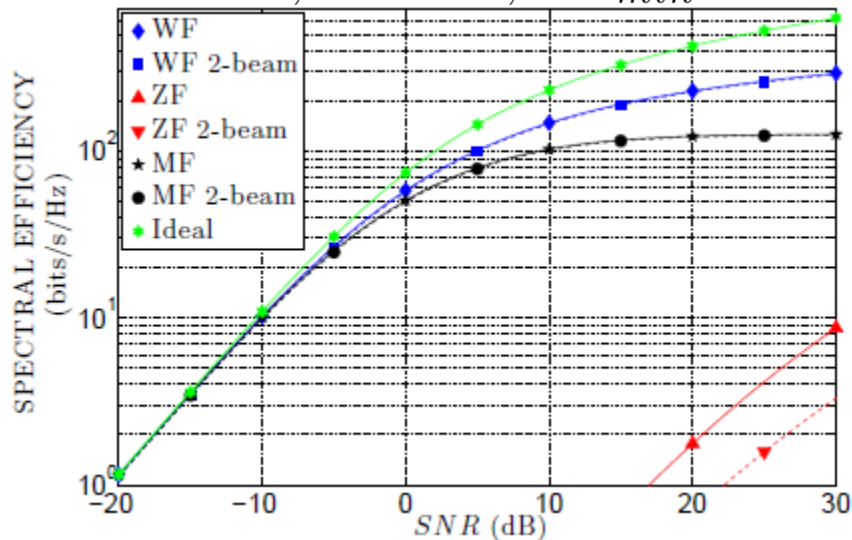


BeamSpace
channel sparsity

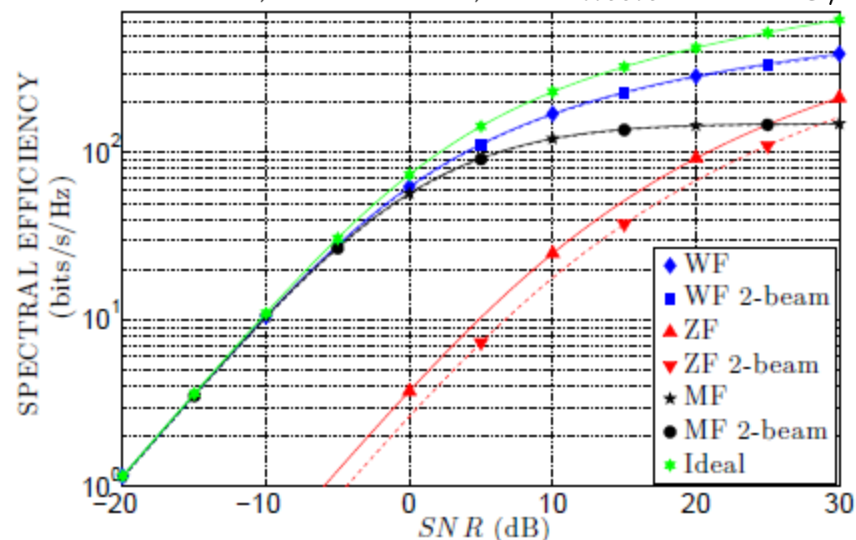


Performance with Linear Precoding

$n = 81, K = 60, \Delta\theta_{min} = 0$



$n = 81, K = 60, \Delta\theta_{min} = \Delta\theta_o/4$

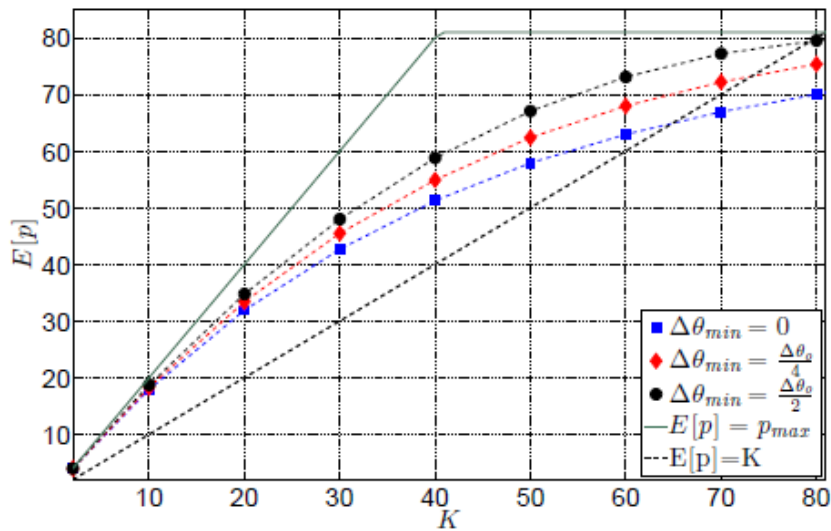


$K = \# \text{ users}$	Spectral Efficiency (bits/s/Hz)	Aggregate rate (Gbps)	Average per-user rate (Gbps)
$K = 20 \Delta\theta_{min} = 0$	134	670	33.5
$K = 20 \Delta\theta_{min} = \Delta\theta_o/4$	159	795	39.8
$K = 40 \Delta\theta_{min} = 0$	192	960	24
$K = 40 \Delta\theta_{min} = \Delta\theta_o/4$	243	1215	30.4
$K = 60 \Delta\theta_{min} = 0$	226	1130	18.8
$K = 60 \Delta\theta_{min} = \Delta\theta_o/4$	283	1415	23.6

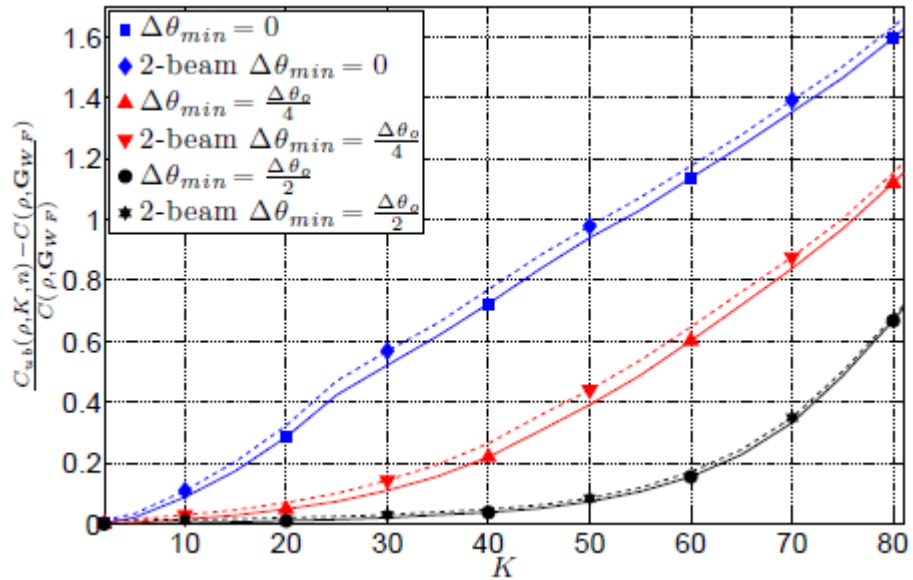
(JB & AS Globecom 2013)



Number of Beams and Capacity Gap



Expected number of active beams
With 2-beam/user sparsity mask



Normalized capacity gap between the upper bound and the MMSE precoder (SNR = 30dB)

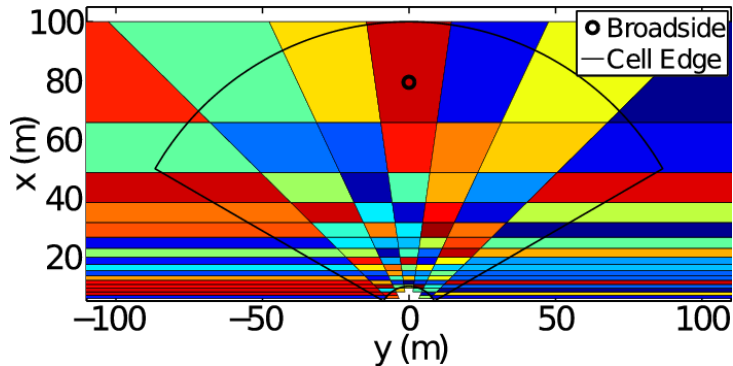
(JB & AS Globecom 2013)

(MMSE precoder:
Joham, Utschick, & Nosssek 2005)

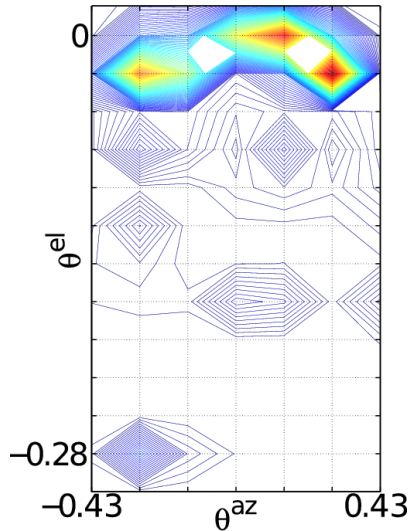
Small-Cell Design: 2D Beam Footprints

$$p = K = 100$$

0.45" x 2.8" antenna @ 80GHz

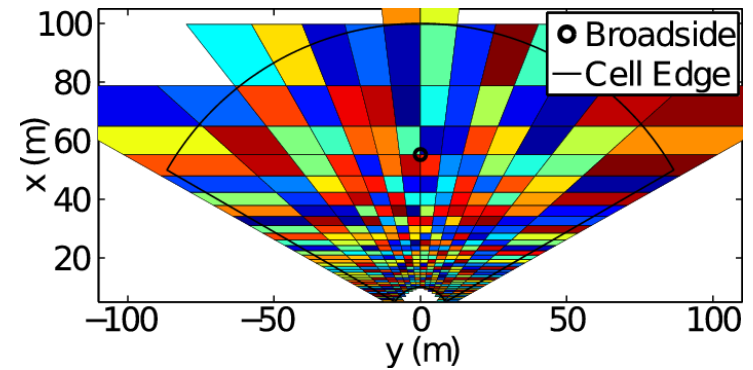


$$n = 273 : (n_{az}, n_{el}) = (7, 39)$$

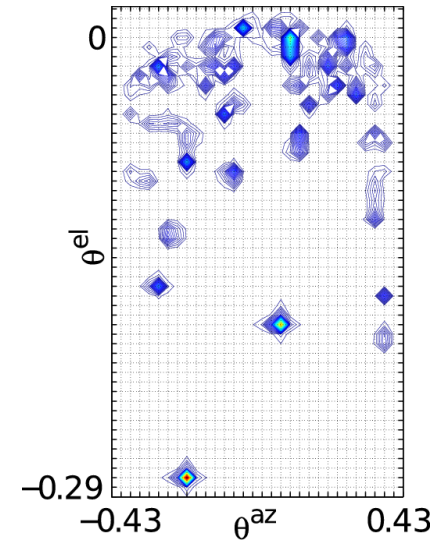
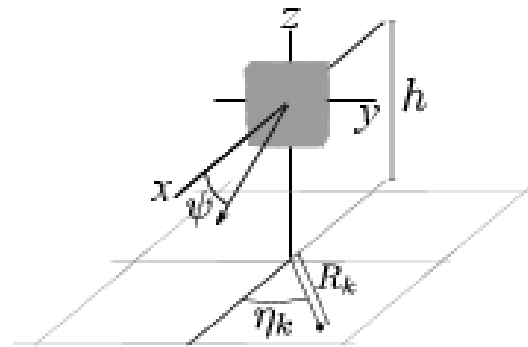


$$p = 16K = 1600$$

2.3" x 12" antenna @ 80GHz

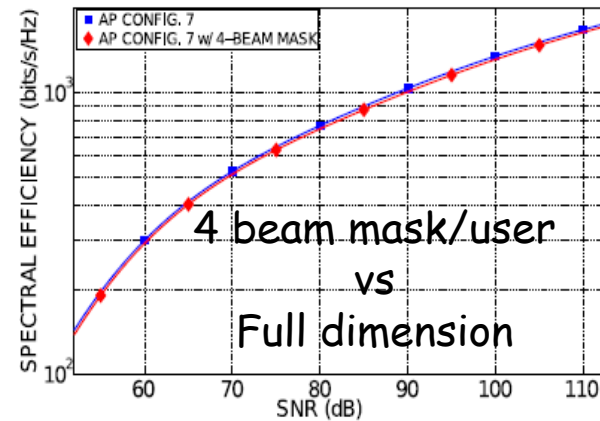
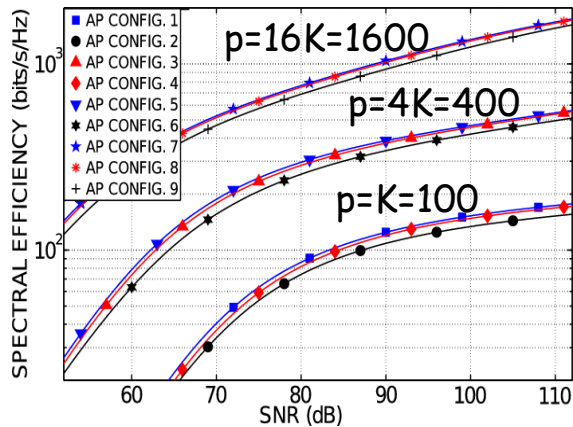


$$n = 5216 : (n_{az}, n_{el}) = (32, 163)$$

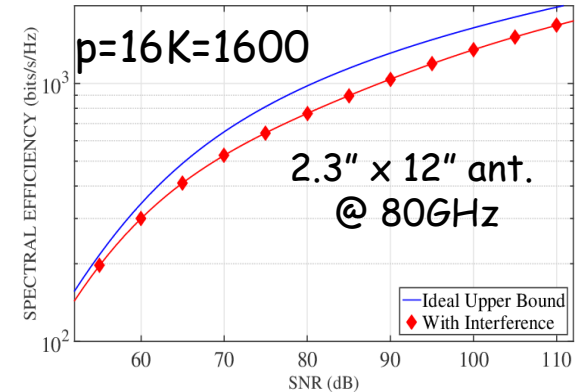
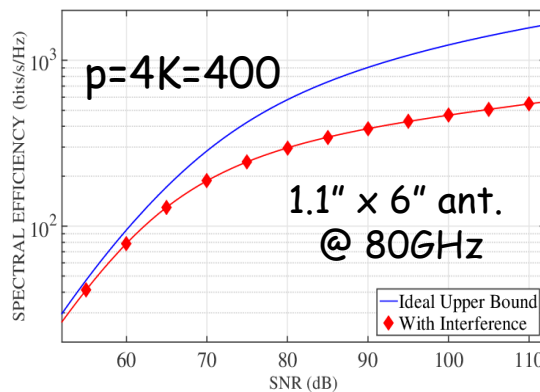
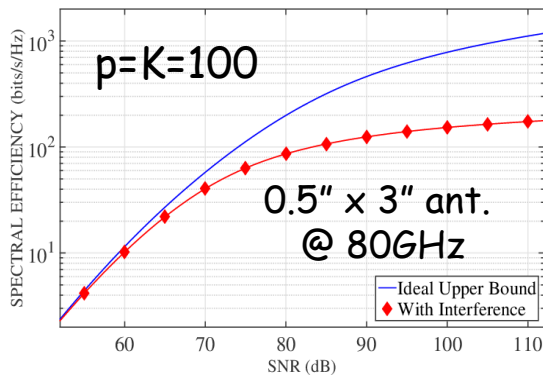


(JB & AS, SPAWC 2014)

Performance of APs with 2D Arrays

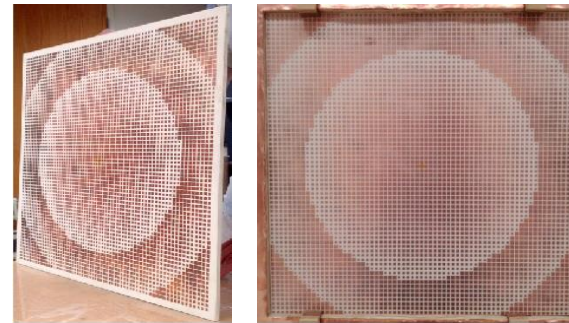
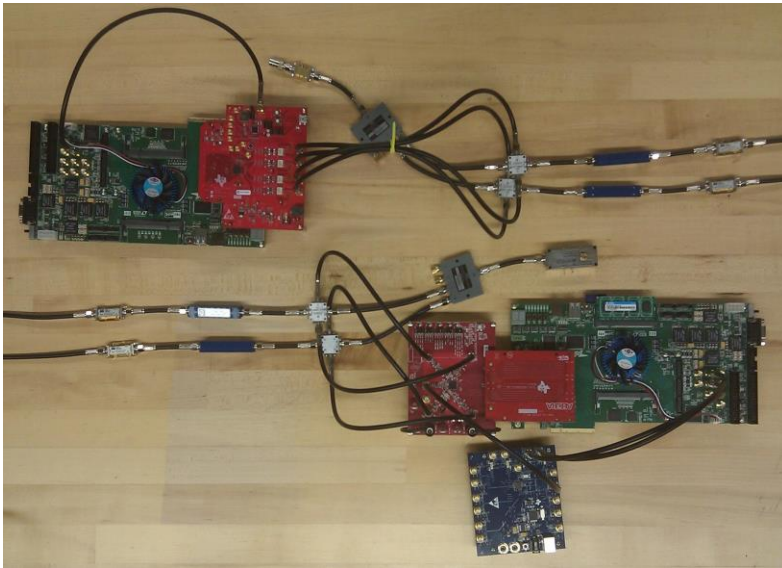


Upperbound vs MMSE precoder performance



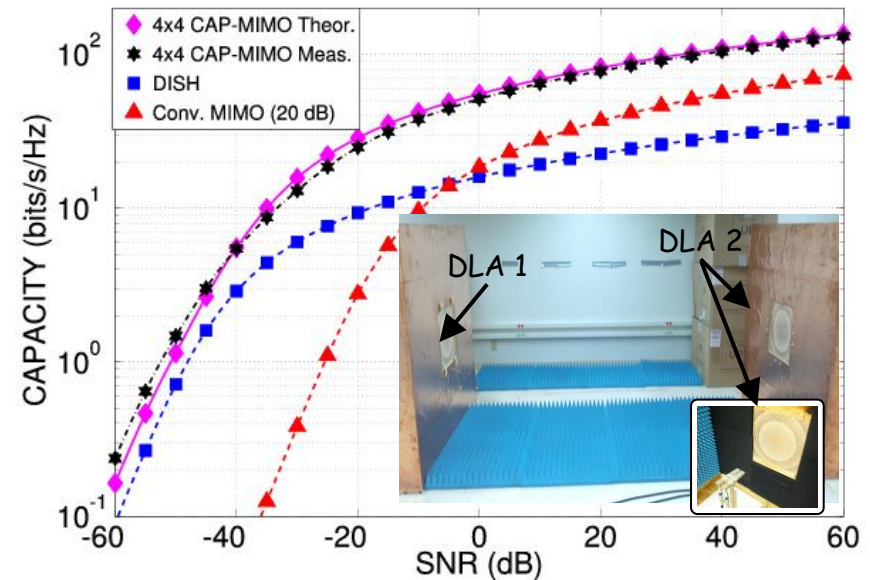
AP config.	$\psi(^{\circ})$	n_{az}	n_{el}	n	p	\tilde{n}_{az}	\tilde{n}_{el}	\tilde{n}	Array Size at 80 GHz
1	7.2	7	39	273	105	7	32	224	$0.45^{33} \times 2.81^{33}$
2	10.3	7	36	245	103	7	29	203	$0.45^{33} \times 2.58^{33}$
3	37	7	34	238	100	7	19	133	$0.45^{33} \times 2.44^{33}$
4	6.4	16	81	1296	408	15	66	990	$1.11^{33} \times 5.91^{33}$
5	10.3	16	76	1216	402	15	61	915	$1.11^{33} \times 5.54^{33}$
6	37	16	70	1120	404	16	39	624	$1.11^{33} \times 5.09^{33}$
7	6	32	163	5216	1610	29	138	4002	$2.29^{33} \times 11.96^{33}$
8	11.1	32	157	5024	1600	29	122	3538	$2.29^{33} \times 11.52^{33}$
9	37	32	151	4832	1614	31	81	2511	$2.29^{33} \times 11^{33}$

10GHz CAP-MIMO Prototype

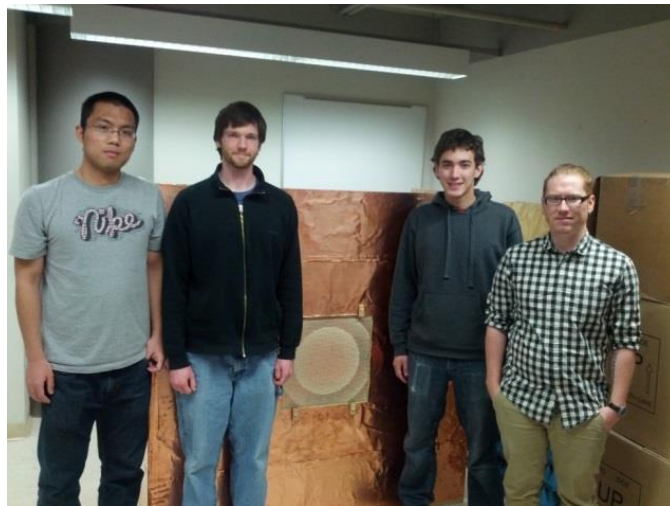


40cm x 40cm
DLA

$R = 10\text{ft}$,
 $n=676$, $p=4$

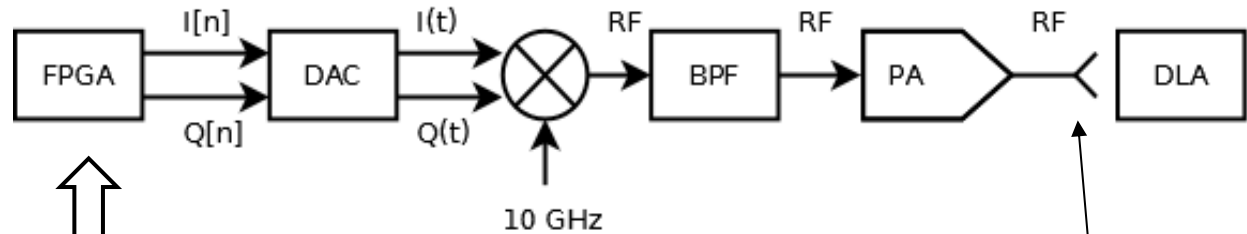


10 GHz prototype theoretical performance:
100 Gigabits/sec (1 GHz BW) at 20dB SNR
Compelling performance gains over state-of-the-art



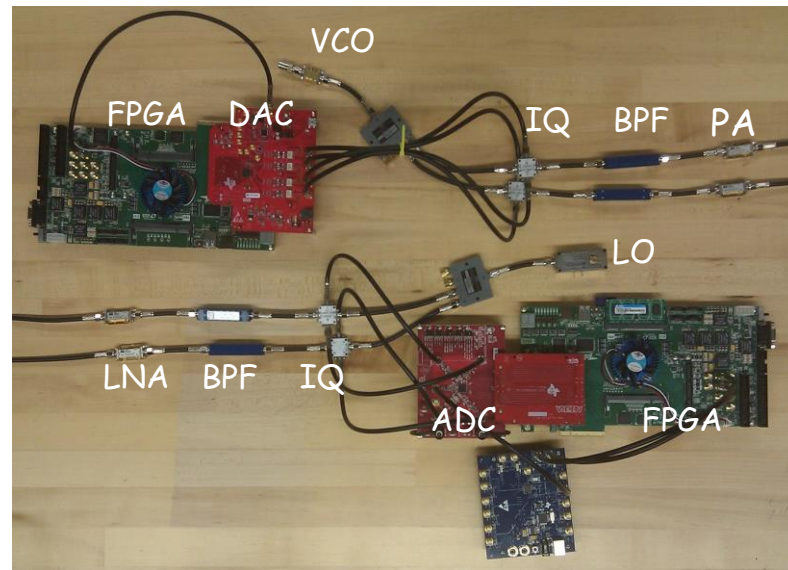
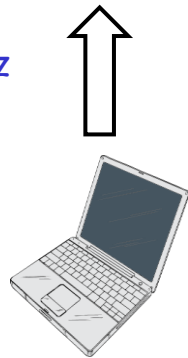
Prototype TX/RX Hardware

Transmitter:



Prototype Specifications

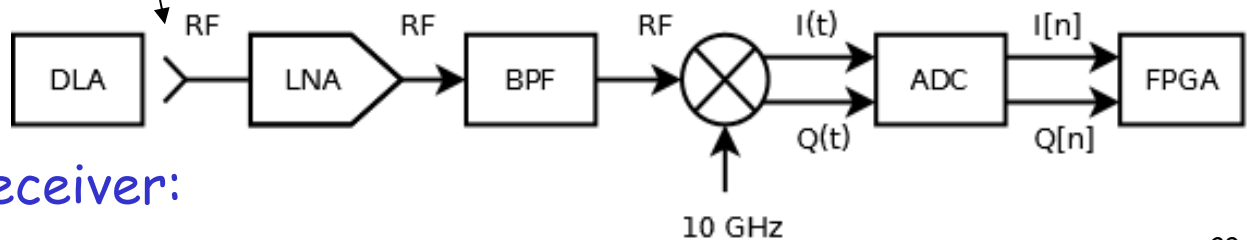
- Operating frequency: 10 GHz
- Up to 4 spatial channels
- 125 MHz symbol rate
- 1 Gigabits/s with 4-QAM
- 8 bits/s/Hz spec. eff.
- TX-RX Clock/Oscillator options
 - Shared clock and oscillator
 - Separate clock or oscillator
 - Separate clock and oscillator



To DLA



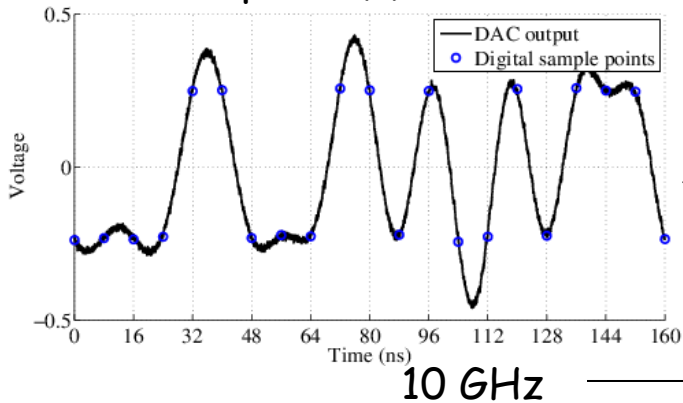
From DLA



Receiver:

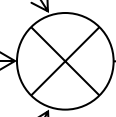
Baseband to Passband

DAC output: $I(t)$

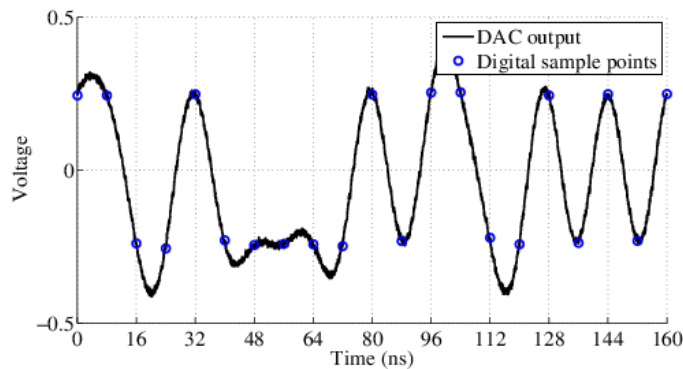


$I(t)$

$u(t)$



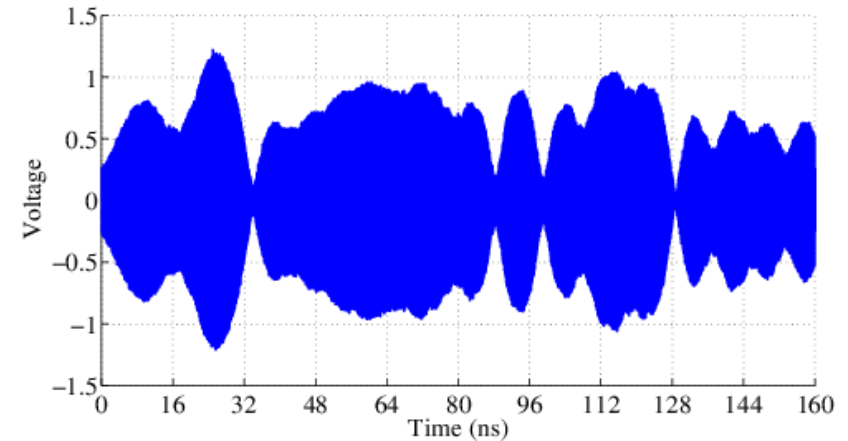
DAC output: $Q(t)$



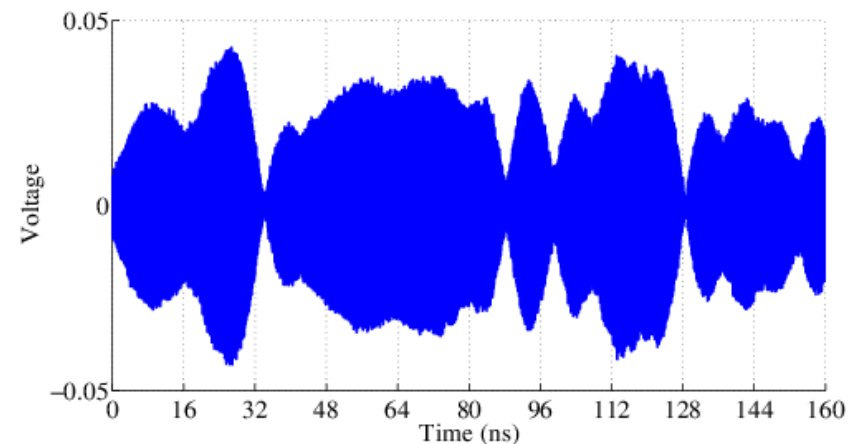
$Q(t)$

$$u(t) = I(t) \cos(2\pi f_c t) - Q(t) \sin(2\pi f_c t)$$

Passband: output of TX PA



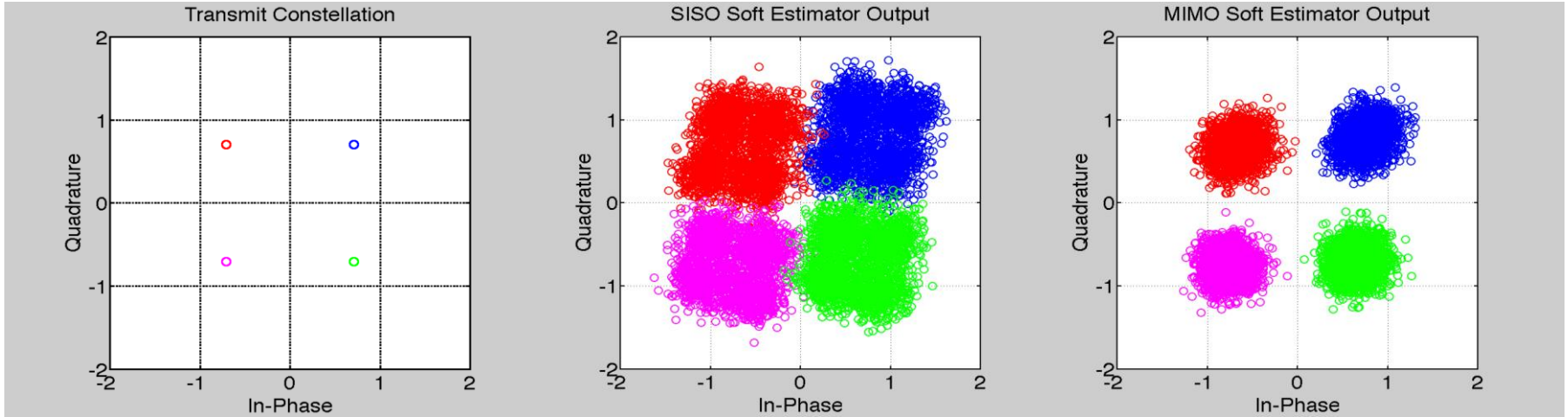
Passband: output of RX LNA



2x2 Spatial Multiplexing Test



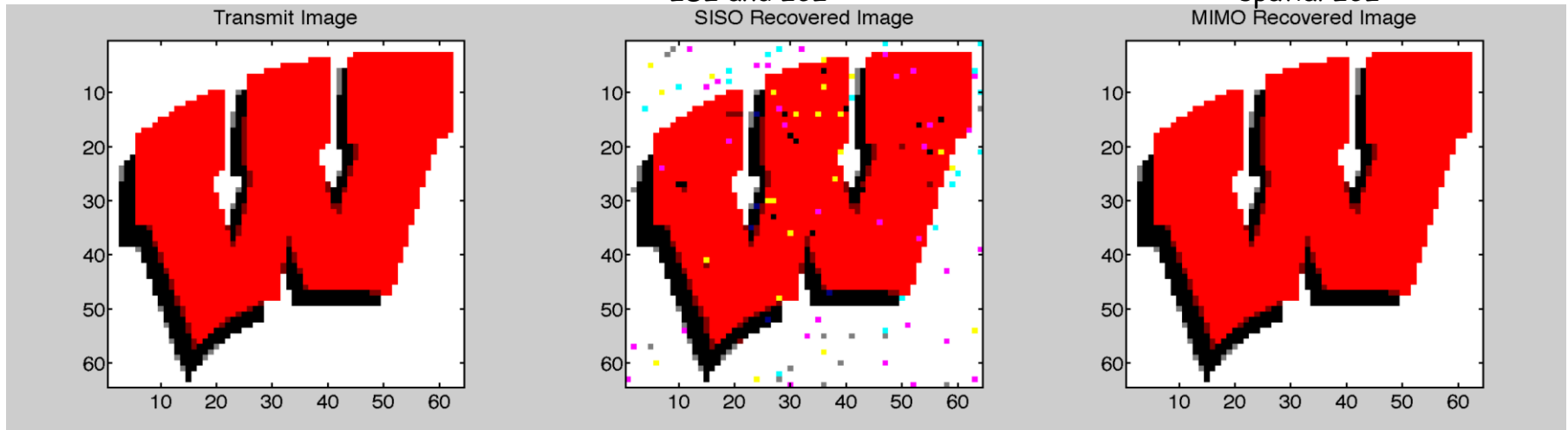
- 2 Spatial Channels
- Shared LO at TX and RX
- 500 Mbps data rate
- Separate TX and RX sample clocks



Transmitted 4-QAM Symbols

Channel 1 received symbols with
ISI and ICI

After MMSE MIMO processing to suppress
spatial ICI



16 kbit test image

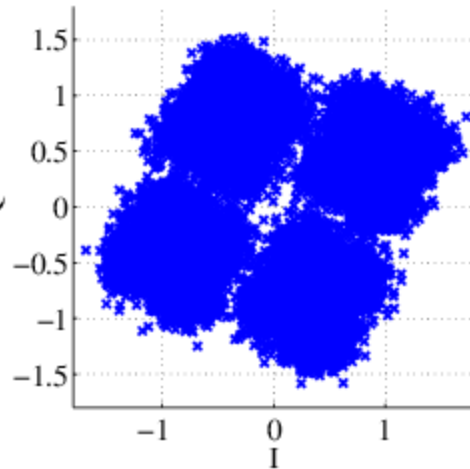
179 bit errors

0 bit errors

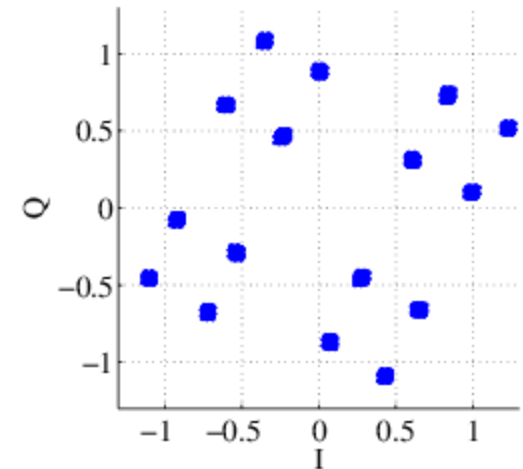
Received Symbols: ICI vs ISI

ISI + ICI

Without
Guard
Interval

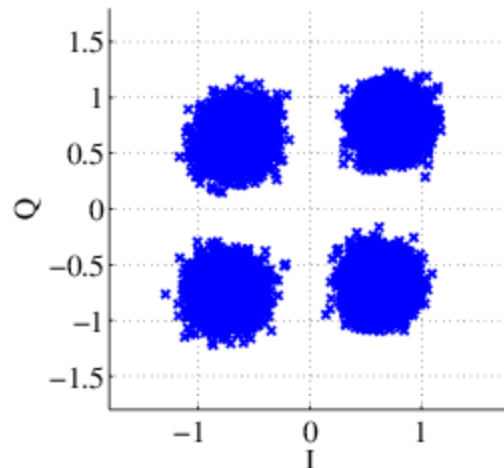


With
Guard
Interval
(ICI only)

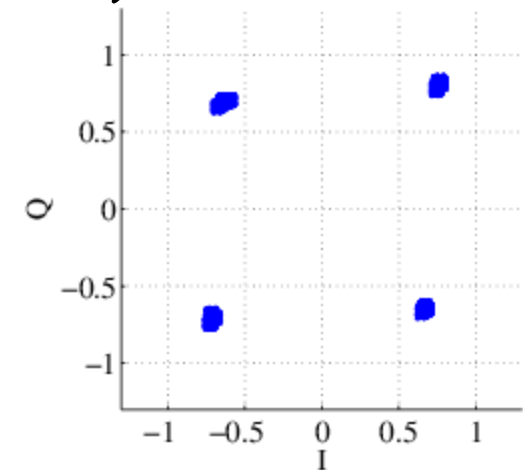


Suppressed ICI (MMSE Spatial Filter)

Without
Guard
Interval
(ISI)

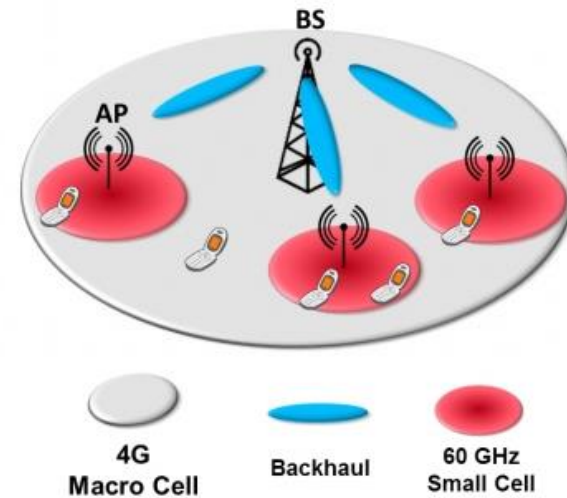
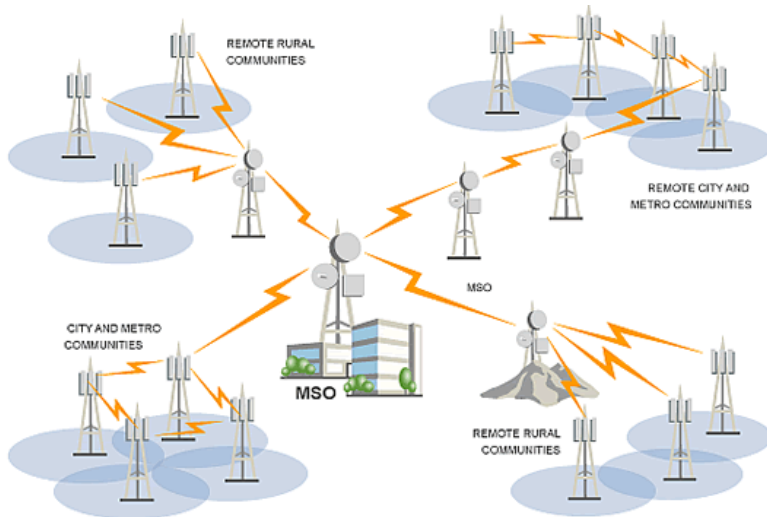
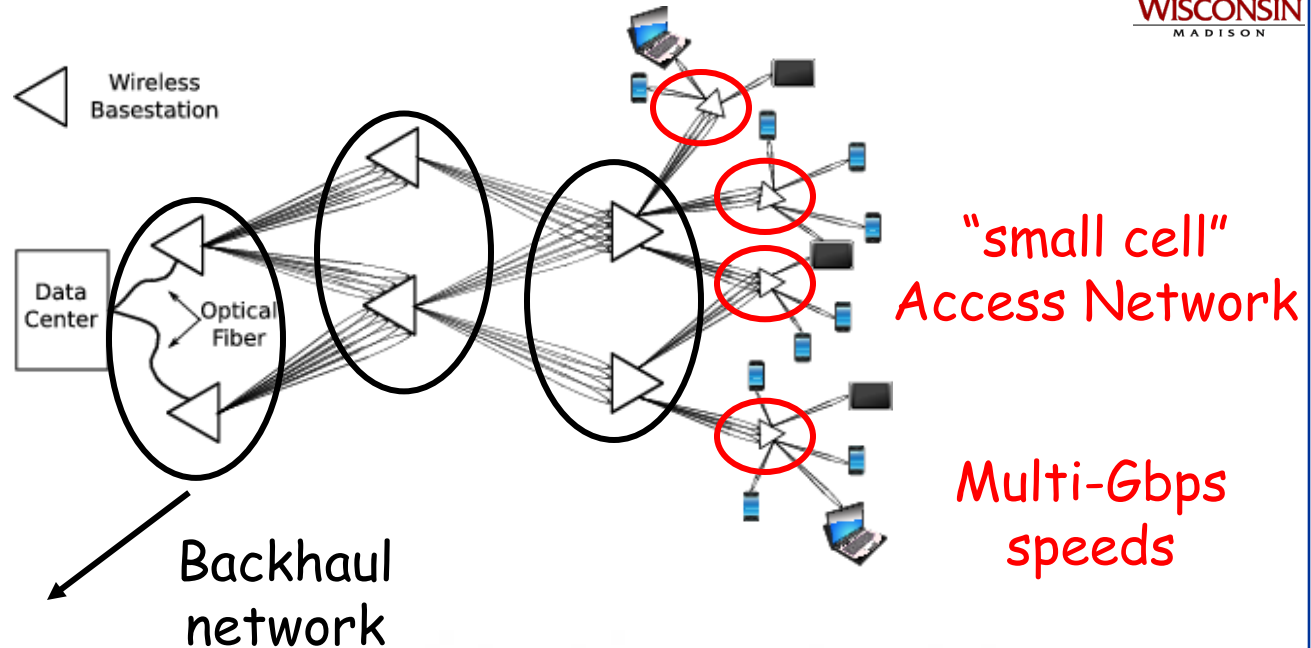
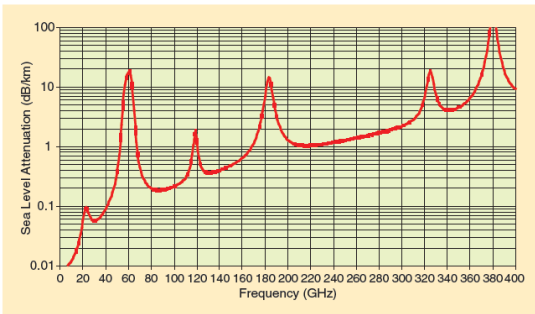


With
Guard
Interval
(suppressed
ISI & ICI)



Outlook: Multi-scale mmW MIMO Networks

Availability:
Atmospheric
absorption

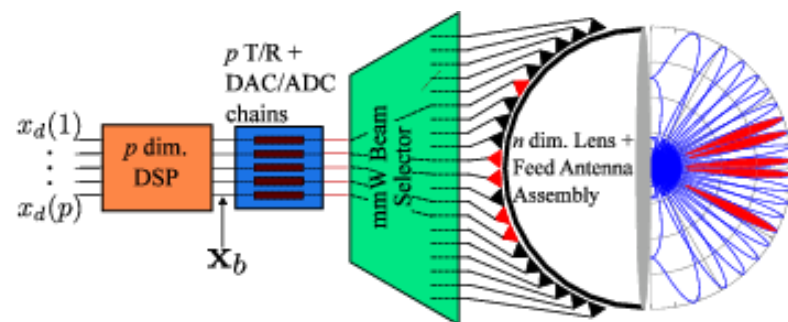


Going Forward

- **Channel Estimation & Discovery**
 - compressed sensing vs thresholding?
 - Analog subspace estimation
- **Spatial Analog-Digital Interface**
 - High symbol rates make DSP very power hungry
 - Move more processing into analog (mmW)?
- **Wideband High-Dimensional MIMO**
 - Need to revisit "narrowband" analysis
 - OFDM, SC, SC-FDMA ... (limited selectivity)
- **Electronic Multi-beam steering**
- **Channel Measurements (true beamspace)**

Conclusion

- **Optimal Beamspace MIMO Communication**
 - Versatile theory & computational framework for system design & optimization
- **CAP-MIMO: practical architecture**
 - spatial A-D interface + DSP
- **Compelling advantages over the state-of-the-art**
 - Capacity/SNR gains
 - Operational functionality
 - **Electronic multi-beam steering & data multiplexing**
- **Timely applications (multi-Gigabits/s speeds)**
 - Long-range wireless backhaul links; Indoor short-range links
 - Smart 5G Basestations: **High-Gain Dense Beamspace Multiplexing**
- **Gen 2 prototype 28GHz - channel meas. + tech. transfer**



Performance-complexity optimization



Relevant Publications

- A. Sayeed, *Deconstructing Multi-antenna Fading Channels*, IEEE Trans. Signal Processing, Oct 2002
- A. Sayeed and N. Behdad, *Continuous Aperture Phased MIMO: Basic Theory and Applications*, Allerton Conference, Sep. 2010.
- J. Brady, N. Behdad, and A. Sayeed, *Beamspace MIMO for Millimeter-Wave Communications: System Architecture, Modeling, Analysis, and Measurements*, IEEE Trans. Antennas & Propagation, July 2013.
- G.-H Song, J. Brady, and A. Sayeed, *Beamspace MIMO Transceivers for Low-Complexity and Near-Optimal Communication at mm-wave Frequencies*, ICASSP 2013
- A. Sayeed and J. Brady, *Beamspace MIMO for High-Dimensional Multiuser Communication at Millimeter-Wave Frequencies*, IEEE Globecom, Dec. 2013.
- J. Brady and A. Sayeed, *Beamspace MU-MIMO for High Density Small Cell Access at Millimeter-Wave Frequencies*, IEEE SPAWC, June 2014.

Thank You!